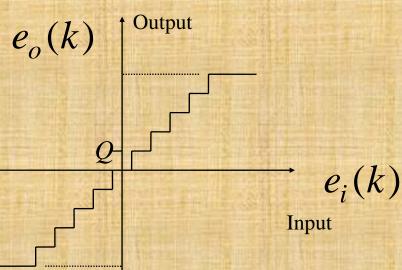
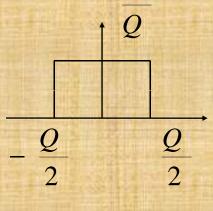
• Finite register lengths and A/D converters cause errors in:-Input quantisation. (i) (ii) Coefficient (or multiplier) quantisation (iii) Products of multiplication or rounded due to truncated machine length

• Quantisation



 $-\frac{Q}{2} \le e_{i,o}(k) \le \frac{Q}{2}$ 

• The pdf for *e* using rounding



• Noise power  $\sigma^2 = \int_{-Q/2}^{Q/2} e^2 p(e) de = E\{e^2\}$ or  $\sigma^2 = \frac{Q^2}{12}$ 

- Let input signal be sinusoidal of unity amplitude. Then total signal power  $P = \frac{1}{2}$
- If b bits used for binary then  $Q = 2/2^{b}$ so that  $\sigma^2 = 2^{-2b}/3$ • Hence  $P/\sigma^2 = \frac{3}{2} \cdot 2^{+2b}$ dB

or SNR = 1.8 + 6b

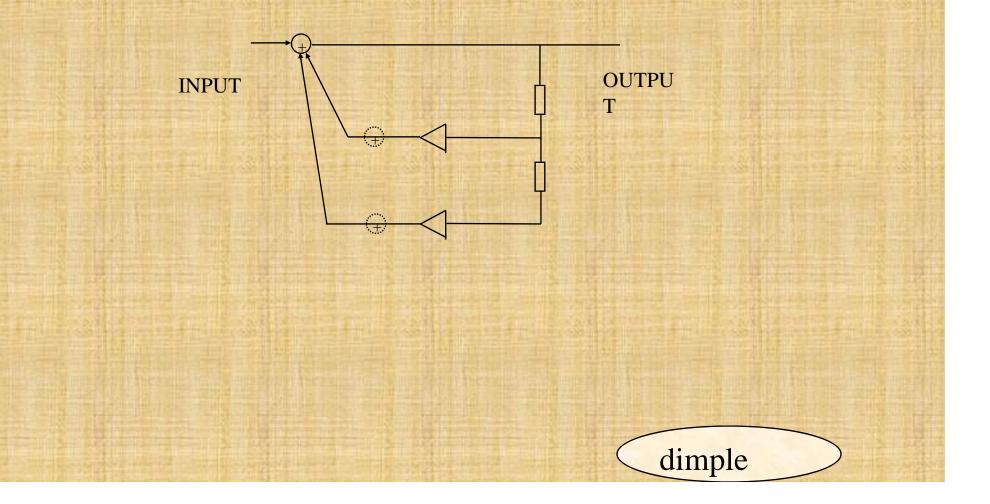
• Consider a simple example of finite precision on the coefficients a,b of second order system with poles  $\rho e^{\pm j\theta}$ 

$$H(z) = \frac{1}{1 - az^{-1} + bz^{-2}}$$

$$H(z) = \frac{1}{1 - 2\rho\cos\theta \cdot z^{-1} + \rho^2 \cdot z^{-2}}$$
  
where  $a = 2\rho\cos\theta \quad b = \rho^2$ 

bit pattern	$2\rho\cos\theta, \rho^2$	ρ
000	0	0
001	0.125	0.354
010	0.25	0.5
011	0.375	0.611
100	0.5	0.707
101	0.625	0.791
110	0.75	0.866
111	0.875	0.935
1.0	1.0	1.0

#### • Finite wordlength computations



## Limit-cycles; "Effective Pole" Model; Deadband

- Observe that for  $H(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$
- instability occurs when  $|b_2| \rightarrow 1$
- i.e. poles are
  - (i) either on unit circle when complex
  - (ii) or one real pole is outside unit circle.
- Instability under the "effective pole" model is considered as follows

• In the time domain with  $H(z) = \frac{Y(z)}{X(z)}$ 

$$y(n) = x(n) - b_1 y(n-1) - b_2 y(n-2)$$

With |b<sub>2</sub>|→1 for instability we have Q[b<sub>2</sub>y(n-2)] indistinguishable from y(n-2)
Where Q[·] is quantisation

- With <u>rounding</u>, therefore we have  $b_2 y(n-2) \pm 0.5$  y(n-2)are indistinguishable (for integers) or  $b_2 y(n-2) \pm 0.5 = y(n-2)$ • Hence  $y(n-2) = \frac{\pm 0.5}{1-b_2}$
- With both positive and negative numbers  $y(n-2) = \frac{\pm 0.5}{1-|b_2|}$

• The range of integers  $\frac{\pm 0.5}{1-|b_2|}$ 

constitutes a set of integers that cannot be individually distinguished as separate or from the asymptotic system behaviour.

• The band of integers  $\left(-\frac{0.5}{1-|b_2|}, +\frac{0.5}{1-|b_2|}\right)$ 

is known as the "<u>deadband</u>".

• In the second order system, under rounding, the output assumes a cyclic set of values of the deadband. This is a <u>limit-cycle</u>.

• Consider the transfer function

$$G(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$$

$$y_k = x_k - b_1 y_{k-1} - b_2 y_{k-2}$$

• if poles are complex then impulse response is given by  $h_k$ 

$$h_k = \frac{\rho^k}{\sin\theta} . \sin[(k+1)\theta]$$

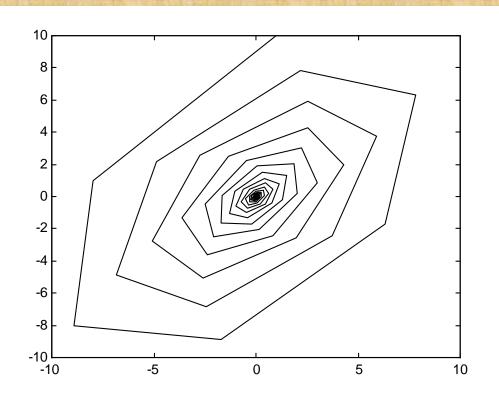
- Where  $\rho = \sqrt{b_2}$   $\theta = \cos^{-1} \left( \frac{-b_1}{2\sqrt{b_2}} \right)$ • If  $b_2 = 1$  then the response is sinusiodal
  - with frequency

$$\omega = \frac{1}{T} \cos^{-1} \left( \frac{-b_1}{2} \right)$$

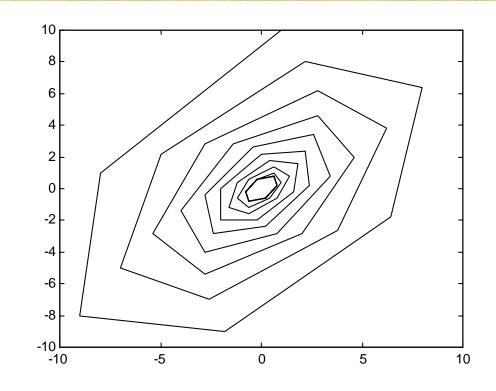
• Thus product quantisation causes instability implying an <u>"effective</u> "  $b_2 = 1$ .

• Consider infinite precision computations for  $y_k = x_k + y_{k-1} - 0.9 y_{k-2}$   $x_0 = 10$ 

 $x_k = 0; k \neq 0$ 



• Now the same operation with integer precision

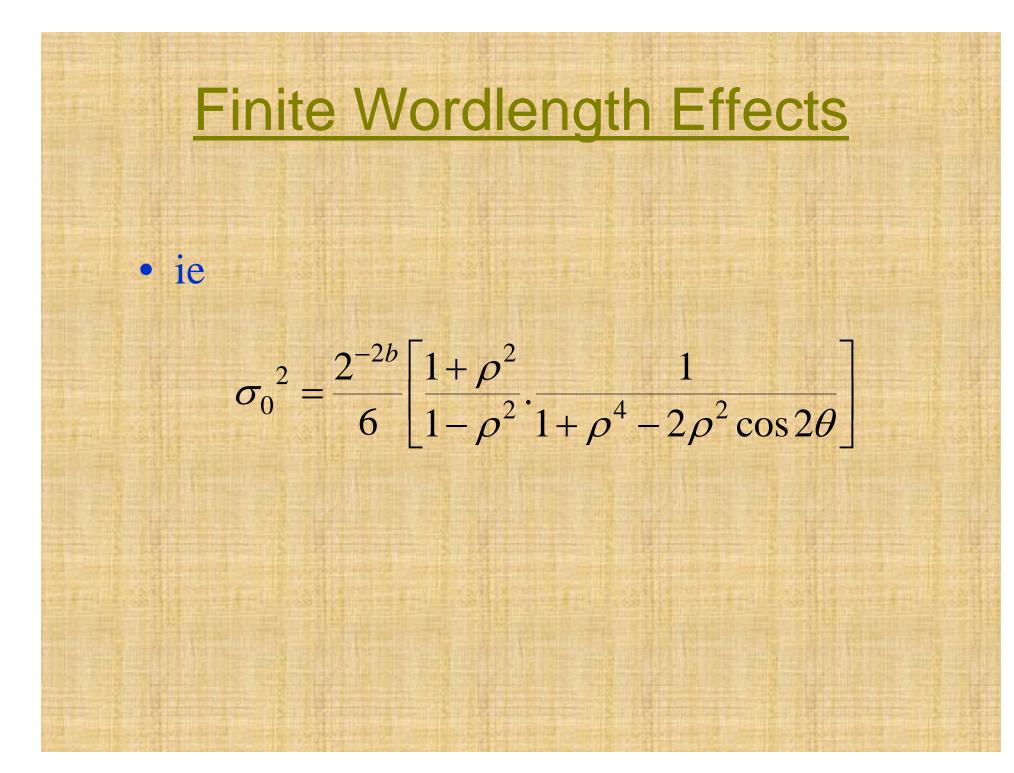


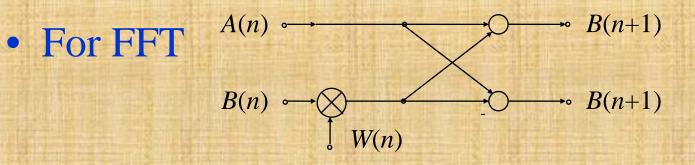
• Notice that with infinite precision the response converges to the origin

• With finite precision the reponse does not converge to the origin but assumes cyclically a set of values –the Limit Cycle

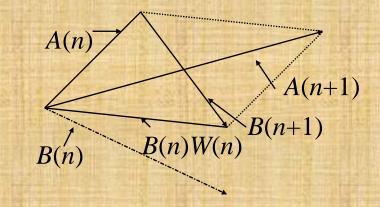
• Assume  $\{e_1(k)\}, \{e_2(k)\}, \dots$  are not correlated, random processes etc.  $\sigma_{0i}^{2} = \sigma_{e}^{2} \sum_{k=0}^{\infty} h_{i}^{2}(k) \quad \sigma_{e}^{2} = \frac{Q^{2}}{12}$ Hence total output noise power  $\sigma_0^2 = \sigma_{01}^2 + \sigma_{02}^2 = 2 \cdot \frac{2^{-2b}}{12} \sum_{k=0}^{\infty} \rho^{2k} \cdot \frac{\sin^2[(k+1)\theta]}{\sin^2\theta}$ • Where  $Q = 2^{-b}$  and

 $h_1(k) = h_2(k) = \rho^k \cdot \frac{\sin[(k+1)\theta]}{\sin\theta}; \ k \ge 0$ 



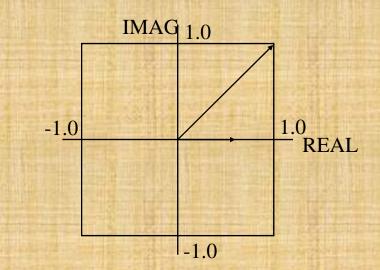


A(n + 1) = A(n) + W(n).B(n)B(n + 1) = A(n) - W(n).B(n)





 $|A(n+1)|^{2} + |B(n+1)|^{2} = 2$  $|A(n+1)|^{2} = 2|A(n)|^{2}$  $|A(n)| = \sqrt{2}|A(n)|$ • AVERAGE GROWTH: 1/2 BIT/PASS



• FFT

 $\begin{aligned} A_x(n+1) &= A_x(n) + B_x(n)C(n) - B_y(n)S(n) \\ &|A_x(n+1)| < |A_x(n)| + |B_x(n)||C(n)| - |B_y(n)||S(n)| \\ &\frac{|A_x(n+1)|}{|A_x(n)|} < 1.0 + |C(n)| - |S(n)| = 2.414 \dots \end{aligned}$ 

• PEAK GROWTH: 1.21.. BITS/PASS

• Linear modelling of product quantisation

$$\begin{array}{c} x(n) \longrightarrow Q[\cdot] \xrightarrow{\widetilde{x}} (n) \\ \end{array}$$

q(n)

• Modelled as

$$x(n) \longrightarrow \widetilde{x}(n) = x(n) + q(n)$$

- For <u>rounding</u> operations q(n) is uniform distributed between  $-\frac{Q}{2}, \frac{Q}{2}$  and where Q is the quantisation step (i.e. in a wordlength of bits with sign magnitude representation or mod 2,  $Q = 2^{-b}$ ).
- A discrete-time system with quantisation at the output of each multiplier may be considered as a multi-input linear system

• Then  $y(n) = \sum_{r=0}^{\infty} x(r).h(n-r) + \sum_{\lambda=1}^{p} \left[ \sum_{r=0}^{\infty} q_{\lambda}(r).h_{\lambda}(n-r) \right]$ • where  $h_{\lambda}(n)$  is the impulse response of the system from  $\lambda$  the output of the multiplier to y(n).

# **Finite Wordlength Effects** • For zero input i.e. $x(n) = 0, \forall n$ we can write $|y(n)| \leq \sum_{\lambda=1}^{p} |\hat{q}_{\lambda}| \cdot \sum_{r=0}^{\infty} |h_{\lambda}(n-r)|$ • where $|\hat{q}_{\lambda}|$ is the maximum of $|q_{\lambda}(r)|, \forall \lambda, r$ which is not more than $\underline{Q}$ 2

vie 
$$|y(n)| \leq \frac{Q}{2} \cdot \sum_{\lambda=1}^{p} \left| \sum_{n=0}^{\infty} |h_{\lambda}(n-r)| \right|$$

• However

$$\sum_{n=0}^{\infty} |h_{\lambda}(n)| \leq \sum_{n=0}^{\infty} |h(n)|$$

• And hence

$$|y(n)| \leq \frac{pQ}{2} \cdot \sum_{n=0}^{\infty} |h(n)|$$

• ie we can estimate the maximum swing at the output from the system parameters and quantisation level